Time-varying fields & Maxwell's equations (TOPIC 5)

beyond the steady state

A time varying magnetic field generates an electric field ("induction"), and a time-varying electric field generates a magnetic field ("displacement current")

The first concept was demonstrated by one of the greatest experimenters of all time, the second one was first introduced by one of the greatest mathematical physicists of history

Why do we study in detail time-dependent fields?

- 1) Only with time-varying fields it is possible to derive the formula of the magnetic field energy
- 2) Only with time-varying fields is the connection between electricity and magnetism evident
- 3) Electric machines (generators, motors and transformers) work with time-varying fields
- 4) Time-varying charges and currents generate electromagnetic radiation: light is a time-varying electric & magnetic field (electromagnetic wave)

Contents of TOPIC 5:

- Electromotive force with static & time-varying magnetic fields: Lorentz-force;
 Faraday's law in integral & differential form; relation Lorentz force & Faraday's law
- Magnetic energy and magnetic forces, applications: generator, motor, transformer
- Time varying electric fields: Ampère-Maxwell's law and displacement currents
- Maxwell's equations, Poynting vector and Poynting's theorem, irradiance
- Electromagnetic waves, refractive index for perfect insulators, Snell's law

Lorentz force law & Electromotive forces

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Lorentz force on moving charges $\vec{F}_{Lorentz} = q\vec{E} + q\vec{v} \times \vec{B}$ Lorentz force "density": $\vec{f}(t,\vec{r}) = \rho(t,\vec{r}) \begin{bmatrix} \vec{E}(t,\vec{r}) + \vec{v}_{drift}(t,\vec{r}) \times \vec{B}(t,\vec{r}) \end{bmatrix}$ (force per unit volume) $= \rho(t,\vec{r})\vec{E}(t,\vec{r}) + \vec{J}(t,\vec{r}) \times \vec{B}(t,\vec{r})$

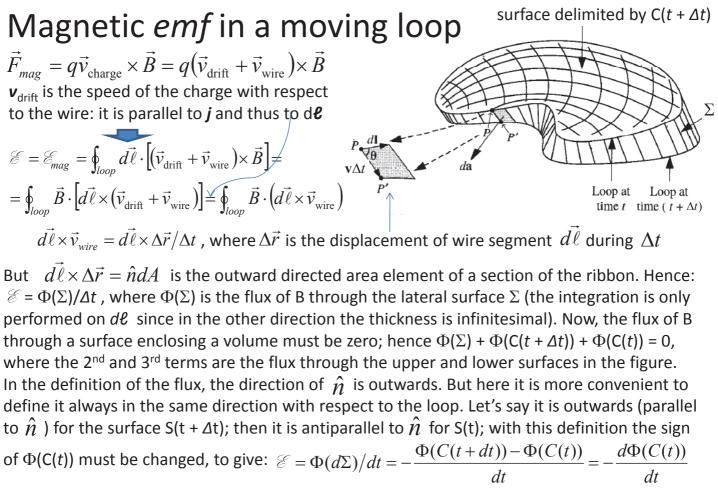
Example: motion in (quasi) uniform B (and E) fields

Polar aurora:
DEF: electromotive force =
$$\mathscr{E} = \int \vec{f} \cdot d\vec{\ell} = \frac{1}{q} \int \vec{F} \cdot d\vec{\ell}$$

(electromotance, emf,
= work per unit charge)
If the applied force
is the Lorentz force: $\mathscr{E} = \int \vec{f} \cdot d\vec{\ell} = \int \frac{\vec{F}_{Lorentz}}{q} \cdot d\vec{\ell} = \int \vec{E} \cdot d\vec{\ell} + \int (\vec{v} \times \vec{B}) \cdot d\vec{\ell} = \mathcal{E}_{elec} + \mathcal{E}_{mag}$
there are 2 ways to generate a current:
- electric electromotance: batteries & fuel cells

- magnetic electromotance: generators

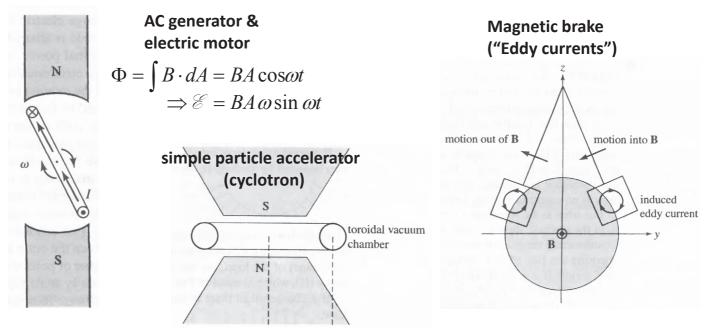
IMPORTANT : the B-field does no work on moving charges, but it does work on magnetic dipoles (and thus on magnetic materials). Moreover, energy must be spent to create magnetic fields.



This magnetic electromotive force (which does not stem from any battery/voltage difference) is non-conservative (the line integral of \mathbf{F}_{mag} around the closed loop is different from zero).

Applications of Lorentz's law and magnetic emf

How can one tell which way a magnetic emf goes? Look at the direction of the magnetic force, or remember Lenz's law: **the current generated by a magnetic** *emf* **always flows so as to oppose the change in external flux**.

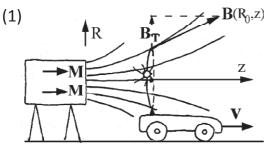


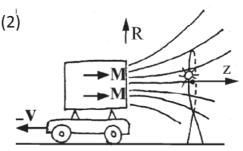
Other applications of the Lorentz force:

- DC electric motor - mass spectrometer

- Hall effect (metal/semiconductor) & polarization of dielectric moving in B-field

Faraday's law: induction





The two situations (1) & (2) are the same, only described by observers moving with relative speed v :

-In (1), the charges in the loop are moving in a static B-field and are subject to the Lorentz force, which acts as electromotive force (topic 4):

$$\mathscr{E} = -\frac{d\Phi_{B}}{dt}$$

-In (2) the same net electromotive force must be present \rightarrow the very existence of the Lorentz force by itself implies a modification of the field equations for **E**: the line integral of E cannot be zero in (2), it must be equal to:

 $\oint \vec{E} \cdot d\vec{\ell} = -\frac{d\Phi_B}{dt}$

In order for the equation of the e.m.f to holds for case (2) no matter what the shape of the loop, it must be:

$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$
 Faraday's law: a changing magnetic field
produces an electric field (& thus an induced e.m.f)

In the most general case, the total e.m.f. in a moving loop which is not connected to a power supply is the sum of two terms, one due to the Lorentz force, and the other due to induction:

$$\mathscr{E} = -\int_{S} \frac{\partial \vec{B}}{\partial t} \cdot d\vec{a} + \oint_{loop} \left(\vec{v} \times \vec{B} \right) \cdot d\vec{\ell} \text{ . The total e.m.f. is always equal to: } \mathscr{E} = -\frac{d\Phi}{dt} = -\frac{d}{dt} \left(\int_{S} \vec{B} \cdot d\vec{a} \right)$$

Implications of Faraday's law

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electromagnetic induction = generation of an E-field by a B-field that varies with time

araday's law
$$\vec{\nabla} \times \vec{E} = -\frac{\partial B}{\partial t}$$

(actually discovered independently by Faraday and Henry in 1830-1831)

Simple application: Find the time-varying **B** field that induces the electric field:

$$\vec{E} = (ax^2y^3\hat{x} + \beta z^4\hat{y})\cos\omega t \qquad \text{Answer:} \quad \vec{B} = (4\beta z^3\hat{x} + 3ax^2y^2\hat{z})\frac{\sin\omega t}{\omega}$$

Since $\vec{B} = \vec{\nabla} \times \vec{A}$, we can write Faraday's law as: $\vec{\nabla} \times \vec{E} = -\frac{\partial(\vec{\nabla} \times \vec{A})}{\partial t}$ or: $\vec{\nabla} \times \left(\vec{E} + \frac{\partial \vec{A}}{\partial t}\right) = 0$

The last equality implies that the field in brackets is a conservative field, hence one can write:

$$\vec{E} + \frac{\partial \vec{A}}{\partial t} = -\vec{\nabla} \mathbf{V}$$
This equation has two extremely important consequences:
1) E is no longer conservative: $\vec{E} = -\vec{\nabla} \mathbf{V} - \frac{\partial \vec{A}}{\partial t}$

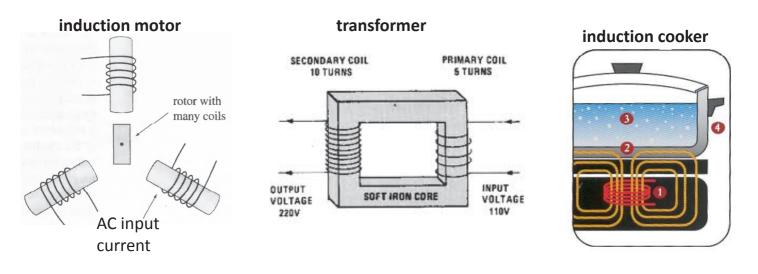
2) In the presence of time-variations, the energy U = qV associated with electrical interactions e.g. inside a battery, goes not only in the creation of **electrical** fields, but also of **magnetic** fields

Technological applications of Faraday's law

$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

Applications of Faraday's discoveries:

- 1) induced emf
- 2) Magnetic energy and self-inductance, inductors
- Foucault (or Eddy) currents: induction stove (heating due to Joule effect and hysteresis loss of ferromagnetic iron pan which also amplifies the magnetic field – you need a special pan for an induction kitchen!)
- 4) induction motor, transformer, ...



Michael Faraday's curriculum vitae

Scientific terms and concepts introduced by M.F. :

Force field, force lines (electric and magnetic lines), ion (anion, cation), voltmeter, electrode (anode/cathode), electrolyte, electrolysis, dielectric, dia- & paramagnetic

Most important scientific discoveries

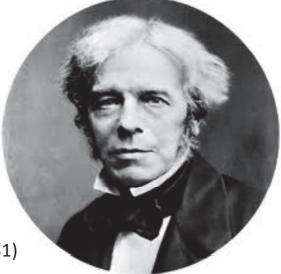
- 1) Faraday's induction law
- 2) Linear dielectric and magnetic response:
 - diamagnetism
 - dielectric constant
- 3) Faraday's laws of electrochemistry
- 4) Faraday rotation (magneto-optic effect)

Inventions

- Electric motor (1821)
- Electric generator and precursor of dynamo (1831)
- precursor of transformer (1831)

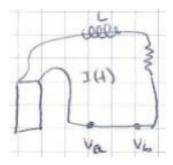
Units named after M.F.:

farad (capacitance) ; faraday (electrochemistry: charge of N_A electrons)



Energy balance in a circuit

Consider a circuit in which the current density is increased from zero to some value J_f . The power source (*e.g.* battery) needs to supply energy to accelerate the electrons; such energy is not only to overcome friction, but also the effect of the opposing Faraday field generated by the increment of J, that is, the corresponding variation of Φ .



Take an element $d\ell$ of the circuit, of cross section da. Let's calculate the work done by the source (battery) to increase the current across $d\ell$ from zero to a finite value I_f . The source maintains a fixed voltage drop across any two points of the circuit. If at a given instant the current through the circuit is I(t), the instantaneous power supplied by the source is equal to :

$$\wp_{battery} = -(V_b - V_a)I(t) = -\Delta V_a$$

If the two points are an infinitesimal distance $d\ell$ away, the voltage drop between them is dV, which is the variation of the scalar potential field across $d\ell$: $dV = \nabla V \cdot d\ell$ Hence, using $Id\ell = \vec{J}_f d\tau$, we get: $d\wp = -(dV)I = -\nabla V \cdot d\ell I = -\nabla V \cdot \vec{J}_f d\tau$

Now, we saw that Faraday's law implies $\vec{E} + \frac{\partial \vec{A}}{\partial t} = -\vec{\nabla}V$. We thus get:

$$d\wp = \vec{E} \cdot \vec{J}_f \, d\tau + \frac{\partial \vec{A}}{\partial t} \cdot \vec{J}_f \, d\tau \Rightarrow \qquad \wp = \int d\tau \, \vec{E} \cdot \vec{J}_f + \int d\tau \, \frac{\partial \vec{A}}{\partial t} \cdot \vec{J}_f$$
(1) mechanical network \wp

(1) mechanical power $\wp_{charges}$

(2) Variation of magnetic energy

The first term is the mechanical power absorbed by the charges, which is finally dissipated as Joule heat. The second one is instead stored as **magnetic energy** as we show in the next slide \rightarrow

Magnetic energy

$$\wp = \int d\tau \, \vec{E} \cdot \vec{J}_{f} + \int d\tau \, \frac{\partial \vec{A}}{\partial t} \cdot \vec{J}_{f}$$
(1) $\wp_{\text{charges}} + \underbrace{\int d\tau \, \frac{\partial \vec{A}}{\partial t} \cdot \vec{J}_{f}}_{(2) \ dU_{\text{mag}}/dt}$

The first term is the mechanical power required to accelerate the charges and keep them moving against the viscous friction drag responsible for the resistance of the wire:

(1)
$$\int d\tau \,\vec{E} \cdot \vec{J}_{f} = \int d\tau \,\vec{E} \cdot \rho_{f} \vec{v}_{drift} = \left(\int dq_{f} \vec{E} \right) \cdot \vec{v}_{drift} = \left(\int d\vec{F} \right) \cdot \vec{v}_{drift} = \vec{F} \cdot \vec{v}_{drift} = \wp_{charges}$$
$$\rho_{f}^{\uparrow} d\tau = dq_{f}$$

Notice that we can extend the integrals to all space, since J_f and \vec{v}_{drift} are zero outside the wire

The 2nd term is the power used up against the induced e.m.f. created by the time-variation of the magnetic field. It is the time-variation of a **magnetic energy** stored in the magnetic field:

(2)
$$\int d\tau \,\frac{\partial \vec{A}}{\partial t} \cdot \vec{J}_{f} = \int d\tau \,\frac{\partial \vec{A}}{\partial t} \cdot \left(\vec{\nabla} \times \vec{H}\right) = \int d\tau \,\vec{\nabla} \cdot \left(\vec{H} \times \frac{\partial \vec{A}}{\partial t}\right) + \int d\tau \,\vec{H} \cdot \left(\vec{\nabla} \times \frac{\partial \vec{A}}{\partial t}\right) = \int d\tau \,\vec{H} \cdot \frac{\partial \vec{B}}{\partial t}$$

In linear media $\vec{B} = \mu_r \mu_0 \vec{H}$. Hence: $= \int \left(\vec{H} \times \frac{\partial \vec{A}}{\partial t} \right) \cdot d\vec{a} = 0$ at ∞

$$(2) = \int d\tau \frac{\vec{B}}{\mu_{\rm r}\mu_0} \cdot \frac{\partial \vec{B}}{\partial t} = \int d\tau \frac{d}{dt} \left(\frac{B^2}{2\mu_{\rm r}\mu_0} \right) = \frac{d}{dt} \int d\tau \frac{B^2}{2\mu_{\rm r}\mu_0} = \frac{d}{dt} \int d\tau \, u_{mag} = \frac{dU_{mag}}{dt} \quad \text{. Therefore:}$$

$$\wp_{source} = \wp_{\rm charges} + \frac{dU_{\rm mag}}{dt} \quad \text{with} \quad U_{mag} = \int d\tau \, u_{mag} \quad \text{and} \quad u_{mag} = \frac{B^2}{2\mu_{\rm r}\mu_0} = \frac{1}{2} \vec{H} \cdot \vec{B}$$

 $U_{\rm mag}$ is the energy stored in the magnetic field (= used to build up the field against induction)

Self-inductance and magnetic energy

The magnetic energy in linear media can also be written as $U_{mag} = \frac{1}{2} \int d\tau \vec{J}_f \cdot \vec{A}$. In fact it is:

$$U_{mag} = \frac{1}{2} \int d\tau \, \vec{H} \cdot \vec{B} = \frac{1}{2} \int d\tau \, \vec{H} \cdot \left(\vec{\nabla} \times \vec{A}\right) = \frac{1}{2} \int d\tau \, \vec{\nabla} \cdot \left(\vec{A} \times \vec{H}\right) + \frac{1}{2} \int d\tau \, \vec{A} \cdot \left(\vec{\nabla} \times \vec{H}\right) = \frac{1}{2} \int d\tau \, \vec{A} \cdot \vec{J}_{f}$$
$$= \frac{1}{2} \int \left(\vec{A} \times \vec{H}\right) \cdot d\vec{a} = 0 \text{ at } \infty$$

According to Biot-Savart's law, the B-field of a loop carrying a uniform current I is: \vec{B}_{loc}

$$p_{p} = \frac{\mu_{0}}{4\pi} \oint \frac{d\tau' \vec{J}(\vec{r}') \times (\vec{r} - \vec{r}')}{|\vec{r} - \vec{r}'|^{3}} = \frac{\mu_{0}}{4\pi} \oint \frac{Id\vec{\ell}' \times (\vec{r} - \vec{r}')}{|\vec{r} - \vec{r}'|^{3}}$$

We see that the field is linearly proportional to I. In the same way, the flux of B through the surface delimited by the loop is also proportional to the current:

$$\Phi_B = \int \vec{B} \cdot d\vec{a} = \frac{\mu_0}{4\pi} I \int \left(\oint \frac{d\vec{\ell}' \times (\vec{r} - \vec{r}')}{|\vec{r} - \vec{r}'|^3} \right) \cdot d\vec{a} = L I \qquad \Phi_B = L I$$

The proportionality coefficient *L* between the flux of B and *I* is the (*self-*)*inductance* of the closed loop. In terms of *L*, Faraday's law becomes: $\mathscr{C} = -\frac{d\Phi_B}{dt} = -L\frac{dI}{dt}$ (this is the formula you use in circuit theory)

Similarly, we find for the magnetic energy of a loop (using $Id\vec{\ell} = \vec{J}_{c}d\tau$):

$$U_{mag} = \frac{1}{2} \int d\tau \, \vec{J}_f \cdot \vec{A} = \frac{1}{2} I \oint d\vec{\ell} \cdot \vec{A} = \frac{1}{2} I \int (\vec{\nabla} \times \vec{A}) \cdot d\vec{a} = \frac{1}{2} I \int \vec{B} \cdot d\vec{a} = \frac{1}{2} I \Phi_B$$

Stokes' theorem
$$\Rightarrow \text{Magnetic energy of a loop:} \quad U_{mag} = \frac{1}{2} I \Phi_B = \frac{1}{2} L I^2 = \frac{\Phi_B^2}{2L}$$

Magnetic vs electric energy formulas

It is interesting to compare the energy formulas we just found with those for electrostatics:

$$u_{mag} = \frac{1}{2} \vec{B} \cdot \vec{H} = \frac{1}{2} \vec{J}_{f} \cdot \vec{A} \qquad u_{el} = \frac{1}{2} \vec{D} \cdot \vec{E} = \frac{1}{2} \rho_{f} V$$

$$L = \frac{\Phi}{I_{f}} \qquad U_{mag} = \frac{1}{2} \Phi I = \frac{1}{2} L I^{2} \qquad C = \frac{Q_{f}}{V} \qquad U_{el} = \frac{1}{2} Q V = \frac{1}{2} C V^{2}$$

$$U = -\vec{m} \cdot \vec{B}_{ext} \qquad U = -\vec{p} \cdot \vec{E}_{ext}$$

*Superconductors: London eq., Meissner effect

Superconducting currents are not really two-dimensional: they are not strictly confined to the surface of the medium but actually penetrate inside it a short distance, called London *penetration depth*. In a normal metal, Ohm's law $J_f = g E$ holds. The fact that **J** (a constant times \mathbf{v}_{drift}) is proportional to **E** is a consequence of electron collisions. But in a superconductor such collisions do not occur; rather, electrons have an accelerated motion: $a_d = \frac{dv_d}{dt} = \frac{eE}{m}$

Since $\frac{\partial J}{dt} = ne \frac{dv_d}{dt}$ (where *n* = density of superconducting electrons) we get, instead of Ohm's law, the London equation: $\frac{\partial J}{\partial t} = \frac{ne^2 E}{m} = \kappa E$ (with $\kappa = \frac{ne^2}{m}$). Taking the curl of both sides and using Faraday's law we obtain: $\frac{\partial (\nabla \times J)}{\partial t} = \kappa \nabla \times E = -\kappa \frac{\partial B}{\partial t}$. Hence apart from a constant that we can discard, we have $\vec{\nabla} \times \vec{J} = -\kappa \vec{B}$. Combining this with Ampère's law for B, $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}$ (valid since a superconductor is nonmagnetic), we get: $ec{
abla} imes ec{
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abla} = -\mu_0 \kappa ec{B}$, which yields, using $ec{
abla} \cdot ec{B} = 0$: B_0 –

$$\nabla^2 \vec{B} = \mu_0 \kappa \vec{B} \equiv \frac{\vec{B}}{\lambda_L^2}$$
 with $\lambda_L = \sqrt{\frac{m}{\mu_0 n e^2}}$ = skin depth

The solution to this equation is (in 1D) : $B(x) = B_0 e^{-x/\lambda}$ λ_L This shows that B decays inside a superconductor in a short distance λ_{I} . Hence B = 0 inside : this is the Meissner effect !

 $B(x) = B_0 e^{-x/\lambda}$ air superconductor

*Condensation energy of a superconductor

We saw in topic 4 that a magnetic field destroys the superconducting state. We can use this to calculate the energy density of the superconducting state at a given temperature T. Take a long cylinder made of a superconducting material, initially in the absence of a magnetic field. As we turn on a weak magnetic field, it fills all space except for the region occupied by the cylinder, since superconducting currents arise that screen the interior of the cylinder from the field. When the applied field reaches the value of the critical field $B_{\rm c}(T)$ at the temperature T, the superconducting state ceases to exist because it is energetically more favorable to have the region it occupies filled by the magnetic field. Hence we can calculate the so-called "condensation energy" density of the superconductor as the equivalent magnetic energy density. The energy density associated with a magnetic field (or, which is the same, with the

the superconducting current density needed to expel it) is $u_{mag} = \frac{1}{2} \vec{B} \cdot \vec{H} = \frac{B^2}{2u_a}$

Here we considered that the normal and superconducting states are basically non-magnetic, so that $\vec{B} = \mu_0 \vec{H}$ (the normal metal is in fact a Pauli paramagnet, but also in this case $\mu_r \cong 1$)

The condensation energy density u_{sc} of the superconductor corresponds to the value of the The condensation energy density u_{SC} of the sequence u_{SC} and $u_{SC} = u_{B_c} = \frac{B_c^2(T)}{2\mu_c}$

The simple formula for the superconducting condensate energy density is actually only valid for a cylinder. For other geometries the amount of energy required to expel the magnetic field turns out to be equal to the magnetic field energy times a geometrical factor.

Mechanical force in an electric circuit

Consider a closed circuit of resistance R and inductance L, connected to a source that provides a constant current I (with I = const, the power transferred to the electrons is dissipated as heat: $\wp_{\text{charges}} = \wp_{\text{Joule}} = R I$). Suppose that we move or deform the circuit in such a way as to vary the inductance L and thus the flux through the circuit. The voltage drop across the resistance R is:

 $\Delta V = \mathcal{E} + \mathcal{E}_{induced} = \mathcal{E} - \frac{d\Phi_B}{dt} = R I$ The e.m.f supplied by the source that maintains the current

constant must overcome the induced e.m.f. due to he change of *L*, as well as the Joule losses Hence:

$$\mathscr{E} = \frac{d\Phi_B}{dt} + R I = \frac{d(L I)}{dt} + R I = \frac{dL}{dt}I + R I$$

On the other hand, by energy conservation, the energy provided by the source must be equal to the change in energy stored in the magnetic field, plus the heat lost by Joule heating, plus the mechanical work *W* to move the coil: $\omega = \mathscr{E} I = \frac{dU_{mag}}{dW_{mag}} + \omega_{mag} + \frac{dW_{mec}}{dW_{mec}}$

$$\wp_{source} = \mathscr{E} I = \frac{d\mathcal{O}_{mag}}{dt} + \wp_{\text{Joule}} + \frac{d\mathcal{W}_{med}}{dt}$$

Assuming that the work is done by a force F across a distance dx, we find:

 $\mathscr{E} Idt = d\left(\frac{1}{2}LI^2\right) + \delta Q_{Joule} + dW_{mec} = \frac{1}{2}I^2 dL + \delta Q_{Joule} + Fdx \qquad \text{(note: } \delta Q_{Joule} \text{ is heat, } \underline{\text{not}} \text{ charge)}$

Plugging the electromotive force found above in the last equation we finally get:

$$\mathscr{E} Idt = dL I^2 + R I^2 dt = \frac{1}{2} I^2 dL + \delta Q_{Joule} + F dx \quad \text{. Since Joule's heat is } \delta Q_{Joule} = R I^2 dt \quad \text{, this gives:}$$

$$E dx = I^2 dI - \frac{1}{2} I^2 dL = +\frac{1}{2} I^2 dL \quad \text{. Hence:} \quad F = \frac{1}{2} I^2 \frac{dL}{dL}$$

This last expression for the force can also be written, using
$$U_{mag} = \frac{\Phi_B^2}{2L}$$
, as: F

$$=-\frac{dU_{mag}}{dx}\Big|_{\Phi_{B}=const}$$

 $|\Phi_{core} = const$

Inductance, energy & force in a magnetic circuit

We saw in topic one that in a magnetic circuit Hopkinson's law applies: $\mathcal{M} = NI = \Re \Phi_{\text{core}}$ Here Φ_{core} is the flux through a cross section of the circuit; for a coil of N turns, the total flux crossing it is $\Phi_B = N\Phi_{core}$. Neglecting the magnetic field outside the circuit, we have:

 $NI = \Re \Phi_{core} = \Re \frac{\Phi_B}{N}$, or: $\Phi_B = \frac{N^2}{\Re}I = LI$. The <u>self-inductance</u> of a coil of *N* turns wrapped around a magnetic circuit of reluctance \Re is thus: $L = \frac{N^2}{\Re}$

The magnetic energy of a magnetic circuit with only one coil around it carrying a current *I* is :

$$U_{mag} = \frac{1}{2}LI^{2} = \frac{1}{2}\frac{N^{2}}{\Re}I^{2} = \frac{1}{2}\frac{M^{2}}{\Re} = \frac{1}{2}\Re\Phi_{core}^{2} = \frac{1}{2}M\Phi_{core}^{2}$$

If there is a gap in the magnetic circuit, there is a force between the opposite poles of the electromagnet across the gap, given by:

$$F = \frac{1}{2}I^2 \frac{dL}{dx} = \frac{1}{2}I^2 N^2 \frac{d}{dx} \left(\frac{1}{\Re}\right) = -\frac{1}{2} \frac{\mathcal{M}^2}{\Re^2} \frac{d\Re}{dx} \text{, that is:} \quad F = -\frac{1}{2} \Phi_{core}^2 \frac{d\Re}{dx}$$

Although we calculated this force in the previous slide assuming a constant current, the force between the poles cannot depend on how the power supply works, but it is always the same for a given pole density, that is, for a given magnetic flux through the core. This is true even if the magnetomotive force is provided by a permanent magnet instead of a coil. Notice that such force can be written, using the expression for the energy given above, as: $F = -\frac{dU_{mag}}{dU_{mag}}$

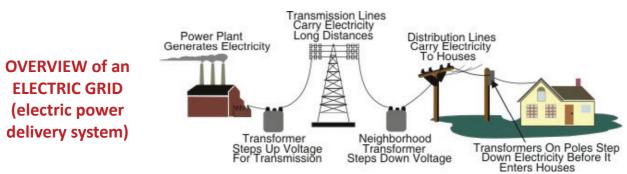
Electric machines & the electric grid

Classification of electric machines:

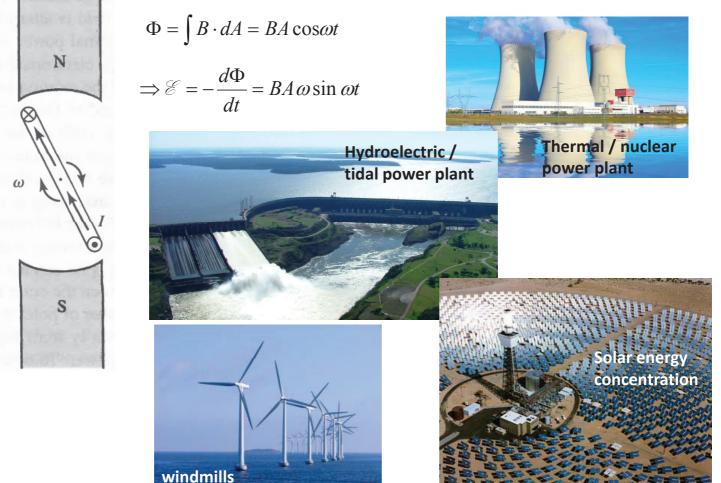
- 1) Machines that convert mechanical energy/work into electric energy: **GENERATORS**
- 2) Machines that convert electric energy into mechanical energy/work: MOTORS
- 3) Machines that inter-convert electric energy: TRANSFORMERS

In cases 1) and 2), the machine contains at least **a coil moving in an external magnetic field**. Whenever this happens, there is:

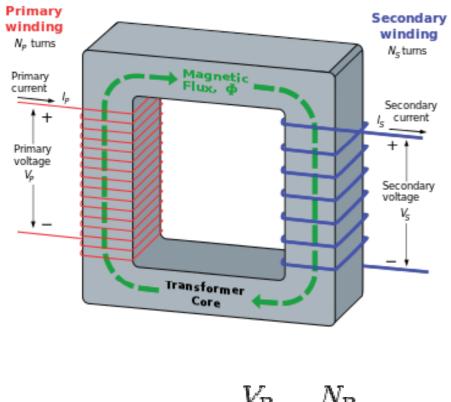
- a) an induced e.m.f./back e.m.f. in the coil, as a result of the time-varying flux;
- b) a mechanical boost/friction force due to the Lorentz force
- It is impossible to have one without the other; however:
- generators are designed in such a way as (also) minimizing the back e.m.f. that would reduce the output voltage;
- motors are designed in such a way as (also) maximizing the current through the loop.
- In both cases part of the task is achieved by ensuring that the magnetic field generated by the current in the loop is negligible with respect to the external field



AC generators (power plants)



(Ideal) transformer



The primary coil consists of $N_{\rm P}$ turns, the secondary one of $N_{\rm S}$ turns. Suppose an AC voltage is applied to the primary coil. By Faraday's law, the induced e.m.f. in the secondary coil is:

$$V_{\rm S} = N_{\rm S} \frac{\mathrm{d}\Phi}{\mathrm{d}t}.$$

Since the same magnetic flux goes through the primary coil, one also must have:

$$V_{\rm P} = N_{\rm P} \frac{\mathrm{d}\Phi}{\mathrm{d}t}.$$

Hence for an ideal transformer

 $rac{V_{
m P}}{V_{
m S}}=rac{N_{
m P}}{N_{
m S}}=a$, where a is the winding turns ratio

A motor has a moving part (**rotor**) and a fixed part (**stator**). There are many types of motors:

Reluctance motor : AC

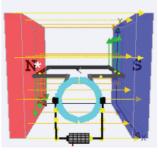
power (usually with different phases) is supplied to the windings (A, B, C) of the stator. The total reluctance of the magnetic circuit made of stator and rotor depends on their relative orientation

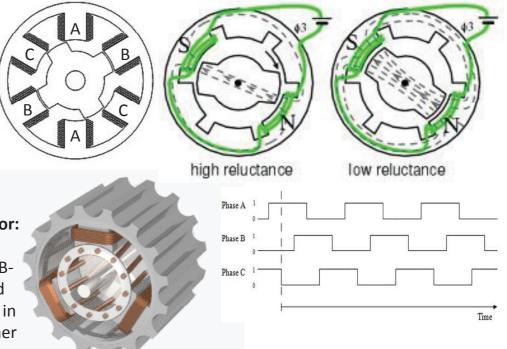
Induction (asynchronous) motor and synchronous motor:

AC power is supplied to the stator windings, producing a Bfield which causes an induced current and a magnetic force in the rotor, which contains either windings or a "squirrel cage"

Motors

(brushed or brushless) **DC motor:** a DC current, whose direction reverses every half-turn, flows through the rotor





Ampère-Maxwell law & displacement current

Ampère's law $\vec{\nabla} \times \vec{H} = \vec{J}_f$ can't hold in general. In fact : a) Taking the divergence of Ampère's law we get: $\vec{\nabla} \cdot \vec{J}_f = \vec{\nabla} \cdot (\vec{\nabla} \times \vec{H}) = 0$. This can't be true

always, since it violates the charge conservation law: $\vec{\nabla} \cdot \vec{J}_f = -\frac{c\rho_f}{\partial t}$ (we see that Ampère's law only holds for *magnetostatics*)

b) Ampère's law in integral form, applied to the charging of a capacitor, gives conflicting results on → different surfaces enclosed by the same loop:

$$\oint \vec{H} \cdot d\vec{\ell} = \int_{S_a} \vec{J}_f \cdot d\vec{a} = I \neq \int_{S_b} \vec{J}_f \cdot d\vec{a} = 0 !!!$$

Maxwell "fixed" Ampère's law by adding an extra term, obtaining:

$$\vec{\nabla} \times \vec{H} = \vec{J}_f + \frac{\partial \vec{D}}{\partial t}$$

Q(t)

a) $\vec{\nabla} \cdot (\vec{\nabla} \times \vec{H}) = \vec{\nabla} \cdot \vec{J}_f + \frac{\partial}{\partial t} (\vec{\nabla} \cdot \vec{D}) = \vec{\nabla} \cdot \vec{J}_f + \frac{\partial}{\partial t} \rho_f \checkmark$

"Displacement current density"

 $\vec{I} = \sigma \vec{E}$

Q(t)

b) Charging of a capacitor at a constant rate (current): $Q_f(t) = I_f$

$$\oint \vec{H}_1 \cdot d\vec{\ell} = \int_{S_1} \vec{J}_f \cdot d\vec{a} + \int_{S_1} \frac{\partial \vec{D}_1}{\partial t} \cdot d\vec{a} = I_f + \varepsilon_0 \int_{S_1} \frac{\partial \vec{E}_1}{\partial t} \cdot d\vec{a} = I_f$$

$$= I_f t$$

$$D_2 = \sigma_f = \frac{Q_f}{S_2} = \frac{I_f t}{S_2} \implies \oint \vec{H}_2 \cdot d\vec{\ell} = \int_{S_2} \vec{J}_f \cdot d\vec{a} + \int_{S_2} \frac{\partial \vec{D}_2}{\partial t} \cdot d\vec{a} = \frac{d}{dt} (D_2 S_2) = I_f \implies H_1 = H_2 \checkmark$$

Displacement current density

displacement
current density $\frac{\partial \vec{D}}{\partial t} = \varepsilon_0 \frac{\partial \vec{E}}{\partial t} + \frac{\partial \vec{P}}{\partial t}$ You saw the 1st term in Fisica 2, it is called
"vacuum" displacement current density

We encountered the 2^{nd} term previously, it is the bound charge current density J_b

- When can the displacement current be neglected?
- 1) In magnetostatics, always (no time-variation of the fields)

2) In general for slow variations (quasi-static approximation):

$$\vec{\nabla} \times \vec{H} = \vec{J}_f + \frac{\partial \vec{D}}{\partial t} \rightarrow$$
 "slow-varying field" $\rightarrow \vec{\nabla} \times \vec{H}(t) = \vec{J}_f(t)$

3) Especially inside metals, at all frequency attainable in electronics (below infrared/visible):

Example: time-varying current in a metal. If $E = E_0 \sin \omega t$, then:

$$\frac{\partial D}{\partial t} = \varepsilon_0 \frac{\partial E}{\partial t} + \frac{\partial P}{\partial t} = \varepsilon_0 \frac{\partial E}{\partial t} = \varepsilon_0 E_0 \omega \cos \omega t \right\} \Rightarrow \frac{|\vec{J}_f|_{\text{max}}}{\left|\frac{\partial \vec{D}}{\partial t}\right|_{\text{max}}} = \frac{gE_0}{E_0 \varepsilon_0 \omega} = \frac{g}{\varepsilon_0 \omega} \approx \frac{10^{18}}{\omega}$$

$$\Rightarrow \frac{I_{displacement}}{I_{free}} = \frac{\varepsilon_0}{g} \omega \approx 10^{-18} \omega \quad \Rightarrow \text{ In a metal, normally: } I_{displacement} << I_{free}$$

 \Rightarrow the slow varying approximation holds in metals at electronic frequencies

Sources of B and H

The Ampère-Maxwell's equation $\vec{\nabla} \times \vec{H} = \vec{J}_f + \frac{\partial \vec{D}}{\partial t}$, besides implying the local form of the charge conservation law, together with Gauss's law for H (or B) it entails that there are **four** types of sources of H and B:

We see that the sources of B are indeed 4 :

- Free currents, \vec{J}_f (1)

- the curl of M, which we know from Ampère's equivalence theorem corresponds to an equivalent current density $\vec{J}_e = \vec{\nabla} \times \vec{M}$ (2)

- the displacement current density $\partial \vec{D} / \partial t$, which is the sum of :

- the "proper" (vacuum) displacement current density, $\varepsilon_0 \frac{\partial E}{\partial t}$ (3) - the bound charge current density, $\frac{\partial \vec{P}}{\partial t}$ (4)

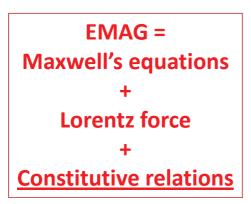
The last term is not surprising since we already know that $\frac{\partial \vec{P}}{\partial t} = \vec{J}_b$ is the current of bound

charges, which is a current of real charges and as such a source of B. Some books define a "total" current density as $\vec{J}_{tot} = \vec{J}_f + \vec{J}_f + \vec{J}_b$, so that one can write an equation similar to the Ampère-Maxwell equation in vacuum you saw in Física 2: $\vec{\nabla} \times \vec{B} = \mu_0 \vec{J}_{tot} + \mu_0 \varepsilon_0 \frac{\partial \vec{E}}{\partial t}$

Summary: axioms of electromagnetism

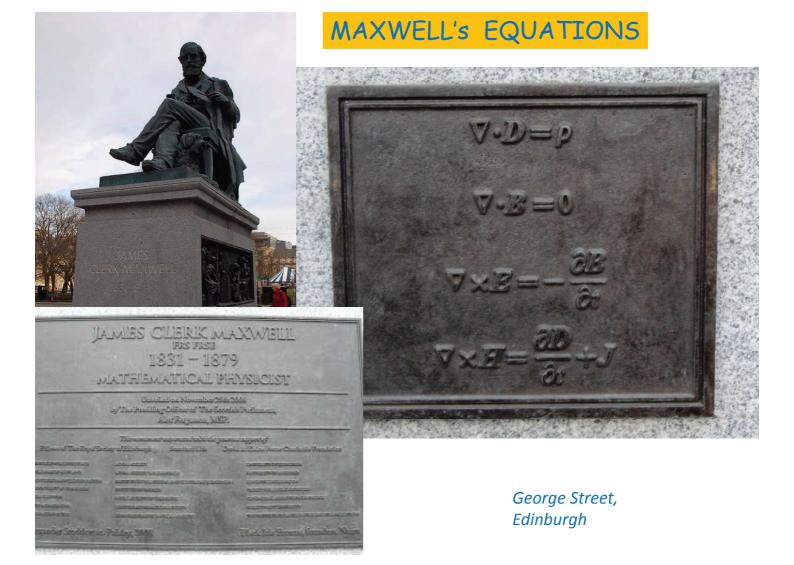
in the absence of material media, we get back the equations of Física 2:

$$\begin{aligned} P_{tot} = \rho_f, \vec{D} = \varepsilon_0 \vec{E} \implies \vec{\nabla} \cdot \vec{E} = \frac{\rho_{tot}}{\varepsilon_0} \\ \vec{\nabla} \cdot \vec{B} = 0 \\ \vec{\nabla} \cdot \vec{E} = -\frac{\partial \vec{B}}{\partial t} \\ \vec{\nabla} \cdot \vec{B} = 0 \\ \vec{\nabla} \cdot \vec{E} = -\frac{\partial \vec{B}}{\partial t} \\ \vec{D} = \varepsilon_0 \vec{E} \\ \vec{B} = \mu_0 \vec{H} \implies \vec{\nabla} \cdot \vec{B} = \mu_0 \left(\vec{J}_f + \varepsilon_0 \frac{\partial \vec{E}}{\partial t} \right) \\ \vec{J}_b = 0, \rho_b \implies \vec{f}_{Lorentz} = \rho_f \vec{E} + \vec{J}_f \times \vec{B} \end{aligned}$$



Some consequences:

- \rightarrow **B** and **E** fields are interrelated !!!
- \rightarrow Poynting's theorem (e.m. energy theorem)
- \rightarrow Existence of e.m. waves !!!
- ightarrow Uniqueness theorem for EMAG



e.m. energy (Poynting's) theorem in vacuum

If we take the dot product of **E** with Ampère-Maxwell's equation $\frac{1}{\mu_0} \vec{\nabla} \times \vec{B} = \varepsilon_0 \frac{\partial \vec{E}}{\partial t}$ we get: $\frac{1}{\mu_0} \vec{E} \cdot (\vec{\nabla} \times \vec{B}) = \varepsilon_0 \vec{E} \cdot \frac{\partial \vec{E}}{\partial t}$. Using the general vector identity $\vec{\nabla} \cdot (\vec{E} \times \vec{B}) = \vec{B} \cdot (\vec{\nabla} \times \vec{E}) - \vec{E} \cdot (\vec{\nabla} \times \vec{B})$, it is : $\varepsilon_0 \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} = \frac{1}{\mu_0} \vec{B} \cdot (\vec{\nabla} \times \vec{E}) - \frac{1}{\mu_0} \vec{\nabla} \cdot (\vec{E} \times \vec{B}) = -\frac{1}{\mu_0} \vec{B} \cdot \frac{\partial \vec{B}}{\partial t} - \vec{\nabla} \cdot \left(\frac{\vec{E} \times \vec{B}}{\mu_0}\right)$ (we also used $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$)

With the definition of Poynting's vector $\vec{S} = \frac{\vec{E} \times \vec{B}}{\mu_0}$, the last equation can be written as: $\varepsilon_0 \vec{E} \cdot \frac{\partial \vec{E}}{\partial t} + \frac{1}{\mu_0} \vec{B} \cdot \frac{\partial \vec{B}}{\partial t} = \frac{\partial}{\partial t} \frac{1}{2} \varepsilon_0 \vec{E}^2 + \frac{\partial}{\partial t} \frac{1}{2\mu_0} \vec{B}^2 = -\vec{\nabla} \cdot \vec{S}$. Since: $\begin{cases} \frac{1}{2} \varepsilon_0 E^2 \rightarrow \text{vacuum electric energy density} \\ \frac{1}{2\mu_0} B^2 \rightarrow \text{vacuum magnetic energy density} \end{cases}$

$$\Rightarrow \frac{\partial}{\partial t} u_{em} + \vec{\nabla} \cdot \vec{S} = 0 \quad \Rightarrow \frac{\text{Poynting's theorem}}{\text{conservation of electromagnetic energy}}$$

In integral form (integrating over the volume) :

$$-\frac{\partial}{\partial t}U_{em} = \int dV \,\vec{\nabla} \cdot \vec{S} = \Phi_{\vec{S}}$$

 \rightarrow the variation of electromagnetic energy (in vacuum) is equal to the flux of Poynting's vector; Poynting's vector takes e.m. energy across the boundary of a volume

Def: time-average of the modulus of Poynting vector = irradiance (\Im) = = (electromagnetic) energy flux per unit area and unit time

Poynting's theorem in linear media

We define the **Poynting's vector** as $\vec{S} = \vec{E} \times \vec{H}$. Taking the divergence of the Poynting vector, we find (using the product rule for derivatives): $\vec{\nabla} \cdot \vec{S} = \vec{\nabla} \cdot (\vec{E} \times \vec{H}) = \vec{H} \cdot (\vec{\nabla} \times \vec{E}) - \vec{E} \cdot (\vec{\nabla} \times \vec{H})$ Using Faraday's law $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$ and Ampère-Maxwell's law $\vec{\nabla} \times \vec{H} = \vec{J}_f + \frac{\partial D}{\partial t}$ this entails: $\vec{\nabla} \cdot \vec{S} = \vec{H} \cdot \left(-\frac{\partial \vec{B}}{\partial t} \right) - \left(\vec{E} \cdot \vec{J}_f + \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \right) = -\vec{E} \cdot \vec{J}_f - \left(\vec{H} \cdot \frac{\partial \vec{B}}{\partial t} + \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \right)$ Since for a linear medium $\vec{E} \cdot \frac{\partial \vec{D}}{\partial t} + \vec{H} \cdot \frac{\partial \vec{B}}{\partial t} = \frac{1}{2} \frac{\partial (\vec{E} \cdot \vec{D})}{\partial t} + \frac{1}{2} \frac{\partial (\vec{H} \cdot \vec{B})}{\partial t} = \frac{\partial (u_{el} + u_{mag})}{\partial t}$ where: $\begin{cases} u_{el} = \frac{1}{2}\vec{E}\cdot\vec{D} \rightarrow \text{electric energy density} \\ u_{mag} = \frac{1}{2}\vec{H}\cdot\vec{B} \rightarrow \text{magnetic energy density} \end{cases} \text{ We then get : } \vec{\nabla}\cdot\vec{S} = -\vec{E}\cdot\vec{J}_f - \frac{\partial}{\partial t}\left(u_{el} + u_{mag}\right) \text{ , or:} \\ \vec{U}_{mag} = \frac{1}{2}\vec{H}\cdot\vec{B} \rightarrow \text{magnetic energy density} \end{cases}$ $-\frac{C\mathcal{U}_{e.m.}}{\partial t} = \vec{E} \cdot \vec{J}_f + \vec{\nabla} \cdot \vec{S}$ In integral form (integrating over a finite volume) : $\frac{\partial U_{em}}{\partial t} = -\wp_{mechanical} - \int d\tau \, \vec{\nabla} \cdot \vec{S}$

 \rightarrow Poynting's theorem (e.m. energy theorem):

The loss of electromagnetic energy in a given volume of space is equal to the flux of Poynting's vector (which carries energy out of the volume) plus the power used to accelerate charges. (the latter is equal to Joule's heat for steady currents inside a conductor)

Electromagnetic waves (light) in vacuum

 $\begin{array}{ccc} \nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t} & \nabla \times \mathbf{B} = -\mu_0 \varepsilon_0 \frac{\partial \mathbf{E}}{\partial t} \\ \nabla \cdot \mathbf{E} = 0 & \nabla \cdot \mathbf{B} = 0 \end{array} \xrightarrow{ \nabla \times \nabla \times \mathbf{E} = -\frac{\partial}{\partial t} \nabla \times \mathbf{B} = -\mu_0 \varepsilon_0 \frac{\partial^2 \mathbf{E}}{\partial t^2} \\ \nabla \times \nabla \times \mathbf{B} = -\mu_0 \varepsilon_0 \frac{\partial^2 \mathbf{E}}{\partial t^2} \end{array} \xrightarrow{ \nabla \times \nabla \times \mathbf{E} = -\mu_0 \varepsilon_0 \frac{\partial^2 \mathbf{E}}{\partial t^2}$ Maxwell's eq. (in vacuum) Using the vector identity: $\nabla \times (\nabla \times \mathbf{V}) = \nabla (\nabla \cdot \mathbf{V}) - \nabla^2 \mathbf{V}$ we get the: e.m. wave equations $\begin{cases} \nabla^2 \vec{E} - \frac{1}{c^2} \frac{\partial^2 \vec{E}}{\partial t^2} = 0 \\ \nabla^2 \vec{B} - \frac{1}{c^2} \frac{\partial^2 \vec{B}}{\partial t^2} = 0 \end{cases}$ Speed of propagation: $\begin{aligned} c = \frac{1}{\sqrt{\varepsilon_0 \mu_0}} = 3 \cdot 10^8 \ m/s \\ c = 1/\sqrt{(8.85 \cdot 10^{-12} s^2 C^2 m^{-3} kg^{-1})(4\pi \cdot 10^{-7} m kg C^{-2})} \end{aligned}$ In 1D: $\frac{d^2y}{dx^2} - \frac{1}{c^2}\frac{d^2y}{dt^2} = 0$ \Rightarrow harmonic solution: $y(x,t) = A\sin(\omega t - kx + \varphi_0)$ $\cos(\omega t - kx + \varphi_0)$ In 3D, **k** and **y** are vectors: $\vec{Y}(\vec{r},t) = \vec{A}\sin(\omega t - \vec{k}\cdot\vec{r} + \varphi_0)$ $\vec{\nabla} \cdot \vec{E} = 0$ $\vec{\nabla} \cdot \vec{B} = 0$ $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \int \vec{P} \cdot \vec{E} \cdot \vec{E} = -\frac{\partial \vec{B}}{\partial t} \int \vec{P} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} = -\frac{\partial \vec{B}}{\partial t} \int \vec{P} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} \cdot \vec{E} = -\frac{\partial \vec{B}}{\partial t} \int \vec{P} \cdot \vec{E} \cdot \vec{E}$ E k \vec{E}

the direction of **E** is called polarization of the e.m. wave (light)

E.m. waves in nonmagnetic linear dielectrics with no free charges nor currents (ideal dielectrics)

Maxwell equations for a nonmagnetic, nonconducting medium ($\mu_r = 1$):

Using the linear constitutive relation for D, we get: $\nabla^2 \mathbf{E} = \mu_0 \mathcal{E}_r \mathcal{E}_0 \frac{\partial \mathbf{E}}{\partial t^2}$

This is the wave equation of waves propagating with speed $v_f = \frac{1}{\sqrt{\varepsilon_r \varepsilon_0 \mu_0}} = \frac{1}{\sqrt{\varepsilon_r}} c = \frac{c}{n}$

where $n = \sqrt{\varepsilon_r}$ = refraction index and $v_f = \frac{c}{n}$ = phase velocity

From Faraday's law we get $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \Rightarrow \vec{k} \times \vec{E}_0 = -\omega \vec{B}_0 \Rightarrow |\vec{B}_0| = \frac{k}{\omega} |\vec{E}_0| = \frac{|\vec{E}_0|}{v_f} = n \frac{|\vec{E}_0|}{c}$

 \mathcal{E}_r is the dielectric constant of electrostatics. If \mathcal{E}_r did not depend on frequency, *n* would be constant. This is only true in the low-frequency limit, while in general $\underline{\varepsilon}$ and *n* depend on $\underline{\omega}$.

Light as electromagnetic energy: irradiance

Energy density (u) associated with E and B fields: $u_e = \frac{1}{2} \varepsilon_0 \varepsilon_r E^2$; $u_m = \frac{1}{2\mu_0\mu_r} B^2$ (with $\mu_r \cong 1$) Since $B = \frac{E}{v_f} = \sqrt{\varepsilon_r} \frac{E}{c}$, it is $u_m = \frac{1}{2\mu_0} B^2 = \frac{n^2}{2\mu_0} \frac{E^2}{c^2} = \frac{1}{2} \varepsilon_r \varepsilon_0 E^2 = u_e \implies u_{e.m.} = u_e + u_m = \varepsilon_r \varepsilon_0 E(t)^2$

Visible light: ω , $v \sim 10^{15}$ Hz , the time variation of **E** can't be measured, nor that of u !!! We can only measure time averages such as that of the energy density $\langle u \rangle$. For a harmonic wave:

 $u_{e.m.} = \varepsilon_r \varepsilon_0 E(t)^2 = \varepsilon_r \varepsilon_0 E_0^2 \sin^2(\omega t) \Longrightarrow \langle u_{e.m.} \rangle = \varepsilon_r \varepsilon_0 E_0^2 \langle \sin^2(\omega t) \rangle = \frac{1}{2} \varepsilon_r \varepsilon_0 E_0^2 \qquad <> = \text{ time average}$

Energy flux divided by area = energy transported by the wave per unit time and area.

$$A = 1 \qquad \overrightarrow{E, B} \qquad \overrightarrow{k} \qquad \overrightarrow{k} \qquad \overrightarrow{Vol} = \ell A = v_f A \Delta t \\ \Delta U_{A=1,\Delta t=1} = \langle u_{em} \rangle \Delta Vol_{A=1,\Delta t=1} = v_f \langle u_{em} \rangle = \Phi_u$$

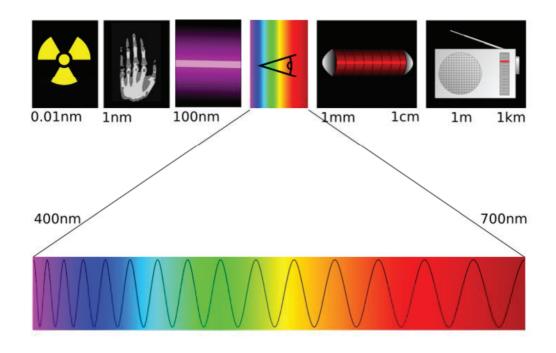
The flux is in the direction of **k**. In vector form : $\vec{S} = \vec{E} \times \vec{H} = \Phi_u \vec{k}$ $\vec{S} =$ **Poynting's vector** In fact, $\vec{E} \times \vec{H}$ is parallel to \vec{k} , and $|\vec{S}(t)| = EH = EB/\mu_0 = E(t)^2/(\mu_0 v_f) = v_f \cdot \varepsilon_r \varepsilon_0 E(t)^2$

<u>The time-averaged energy crossing a surface per unit time and unit surface</u> is called **irradiance**, and it equals the time average value of the modulus of \vec{S} . For a monochromatic e.m. wave:

$$\Im = \langle S \rangle = v_f \langle u_{e.m.} \rangle = \frac{1}{2\mu_0 v_f} E_0^2$$

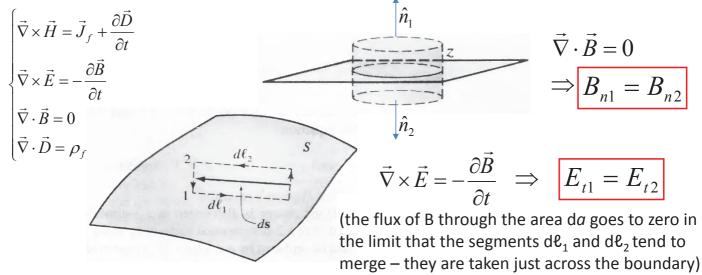
NOTE: light (e.m. wave) carries not only energy, but also linear and angular momentum

Types of light: spectral ranges



harmonic waves travelling at speed c (or c/n): $\lambda f = c$ (or c/n) f is a constant $\Rightarrow \lambda$ (inside dielectric) = λ (vacuum)/n = λ_0 /n Wave frequency (and vacuum wavelength) is associated with **COLOR**

Boundary conditions in optics



Normally there is neither free surface charge nor free surface current on the interface. In such case the four boundary conditions that must be satisfied (for linear media) are:

 $\epsilon_1 E_{1\perp} = \epsilon_2 E_{2\perp}$ and $B_{1\perp} = B_{2\perp}$,

 $\mathbf{E}_{1\parallel} = \mathbf{E}_{2\parallel}$ and $\mathbf{B}_{1\parallel}/\mu_1 = \mathbf{B}_{2\parallel}/\mu_2$

Usually in optics one deals with nonmagnetic media. Even in the case of soft ferromagnets, the magnetization cannot follow applied fields at frequencies higher than few MHz, so that the contribution of M to B at IR or optical frequencies is negligible. Hence the boundary conditions for article fields because simply.

for optical fields become simply:

B = const , E_{//} = const ,
$$\epsilon_1 E_{1\perp} = \epsilon_2 E_{2\perp}$$

Reflection & refraction at a planar interface

Boundary between media: light is partially reflected (R) and partially transmitted (T) (that is, refracted). The E (or B) field on the left-hand side of the boundary is the sum of \mathbf{E}_{I} and \mathbf{E}_{R} , on the other side it is equal to ${\bf E}_{\rm T}$ Boundary conditions in optics: $\mathbf{B} = \text{const}, E_{//} = \text{const}$

 \rightarrow There exists a fixed relation between the fields at all points \vec{r} of the boundary. For a monochromatic incident wave, this

incident implies that the three waves (incident, reflected & transmitted) must have equal total phase : $\vec{k}_I \cdot \vec{r} - \omega_I t = \vec{k}_R \cdot \vec{r} - \omega_R t = \vec{k}_T \cdot \vec{r} - \omega_T t$ Setting the origin $\vec{r} = 0$ at a point on the boundary, at the origin $-\omega_I t = -\omega_R t = -\omega_T t$, hence

 \Rightarrow law of specular reflection

 $\omega_I = \omega_R = \omega_T$ (consistent with the definition of photon and with energy conservation) For t = 0: $\vec{k}_I \cdot \vec{r} = \vec{k}_R \cdot \vec{r} = \vec{k}_T \cdot \vec{r} \implies \vec{k}_I$, \vec{k}_R and \vec{k}_T are coplanar

Taking \vec{r} to be coplanar with the wave vectors we get: $\vec{k}_I \cdot \vec{r} = \vec{k}_R \cdot \vec{r} \Longrightarrow k_I r \sin \theta_I = k_R r \sin \theta_R$ This implies: $k_I \sin \theta_I = k_R \sin \theta_R = k_T \sin \theta_T$

 $\theta_R = \theta_I$ (incident and reflected waves propagate in the same medium of refractive index $n_I \rightarrow$ they have the same value of $|\mathbf{k}|$)

 \Rightarrow Snell's law $n_T \sin \theta_T = n_I \sin \theta_I$ (the transmitted wave propagates in medium with $n_T \neq n_I$ with same frequency as the incident wave $\rightarrow \lambda$ changes)

Reflectance for normal incidence

Boundary conditions in optics: $\mathbf{B} = \text{const}, E_{//} = \text{const}$. For light impinging at normal incidence on a planar boundary this implies:

$$E_{0I} + E_{0R} = E_{0T}$$

 $B_{0I} + B_{0R} = B_{0T}$

From Faraday's law:

$$\vec{\nabla} \times \vec{E} = -\frac{\partial B}{\partial t} \Rightarrow kE_0 = \omega B_0 \Rightarrow B_0 = \frac{E_0}{\omega/k} = n\frac{E_0}{c}$$

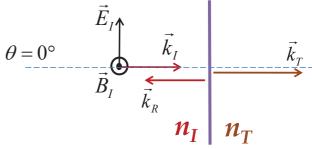
$$\implies \begin{cases} E_{0I} + E_{0R} = E_{0T} \\ n_I E_{0I} \bigoplus n_I E_{0R} = n_T E_{0T} \end{cases}$$
The minus sign takes into account the change in relative orientation of E & B in the reflected wave (as $\vec{k}_R = -\vec{k}_I$)

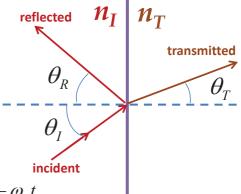
Solving for the transmitted and reflected field amplitudes:

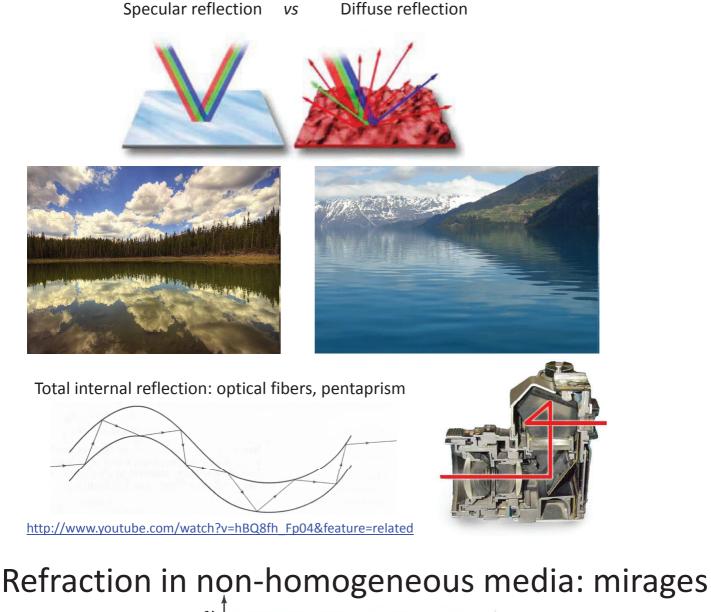
$$\begin{cases} E_{0T} = \frac{2n_{I}}{n_{I} + n_{T}} E_{0I} & \text{We thus get:} \\ E_{0R} = \frac{n_{I} - n_{T}}{n_{I} + n_{T}} E_{0I} & \frac{E_{0R}}{E_{0I}} = \frac{n_{I} - n_{T}}{n_{I} + n_{T}} \Rightarrow \\ R = \frac{\mathfrak{I}_{R}}{\mathfrak{I}_{I}} = \frac{\left|E_{0R}\right|^{2}}{\left|E_{0I}\right|^{2}} = \left|\frac{n_{I} - n_{T}}{n_{I} + n_{T}}\right|^{2} \\ \text{REFLECTANCE} \end{cases}$$

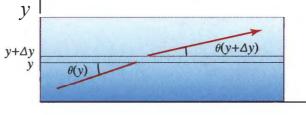
The transmittance T is defined as $T = \frac{\Im_T}{\Im_T}$. For energy conservation, it must be: R + T = 1

NOTE: reflectance and transmittance are the same if light impinges from one side or the other. For the air-glass interface (e.g. at a clean window), R is about 4% at normal incidence, so a window reflects at least 8% of the incoming light (R increases with increasing incidence angle)

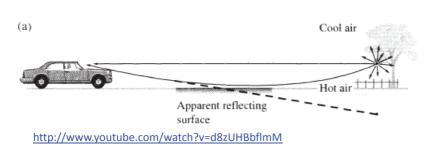








Consider an inhomogeneous medium whose refractive index varies with the vertical coordinate y. A light ray undergoes refraction wherever there is a change of refractive index. Due to Snell's law, since n is varying continuously with y, a light wave travelling through this medium undergoes a continuous refraction \rightarrow mirages:





This effect leads to the phenomenon of mirages, and is also exploited commercially to achieve so-called graded-index materials (GRIN), used for lenses, waveguides and fibers.

GRIN waveguide

